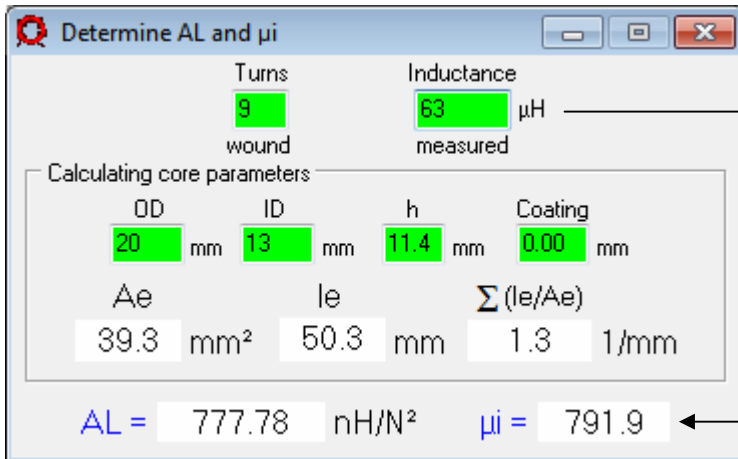


MFJ 854 External Current Probe

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This probe is very easy to use and replaces the clamp-on ferrite which is mounted on the MFJ housing. The ferrite is larger and allows checking the current on coaxes up to 0.55 in. of diameter. Once the probe has been recalibrated, (1 Amp @ 7 MHz) all current scales are functional as before.

Existing clamp-on ferrite



The screenshot shows a software window titled "Determine AL and μ_i ". It contains the following data:

Turns wound	Inductance measured
9	63 μH

Calculating core parameters:

OD	ID	h	Coating
20 mm	13 mm	11.4 mm	0.00 mm

A_e	l_e	$\sum (l_e/A_e)$
39.3 mm ²	50.3 mm	1.3 1/mm

AL = 777.78 nH/N² μ_i = 791.9

Annotations: An arrow points from the measured inductance of 63 μH to the text "Measured: 50 uH Yields: Fc ~ 190 KHz". Another arrow points from the calculated permeability $\mu_i = 791.9$ to the text "Material #43".

The permeability (μ) was determined using Mini Ring Core Calculator s/w plus core dimensions and measured inductance.
 $\mu \sim 800$ which is material # 43

New ferrite used is larger to provide clearance for larger cables

Determine AL and μ_i

Turns wound: 9

Inductance measured: 81 μH

Calculating core parameters:

OD	ID	h	Coating
25.4 mm	15.5 mm	12.7 mm	0.00 mm

A_e : 61.6 mm² l_e : 61.7 mm $\Sigma(l_e/A_e)$: 1.0 1/mm

AL = 1000.00 nH/N² μ_i = 797.1

→ $F_c \sim 118 \text{ KHz}$

Determine AL and μ_i

Turns wound: 10

Inductance measured: 100 μH

Calculating core parameters:

OD	ID	h	Coating
25.4 mm	15.5 mm	12.7 mm	0.00 mm

A_e : 61.6 mm² l_e : 61.7 mm $\Sigma(l_e/A_e)$: 1.0 1/mm

AL = 1000.00 nH/N² μ_i = 797.1

→ $F_c \sim 95 \text{ KHz}$

← MATÉRIEL #43

Hinged (Snap On) Free Hanging Ferrite Core 90Ohm @ 100MHz ID 0.610" Dia (15.49mm) OD 1.360" W x 0.591" H (34.54mm x 15.01mm)

Length 0.835" (21.21mm) #43 material 0443806406 1934-1209-ND \$2.84

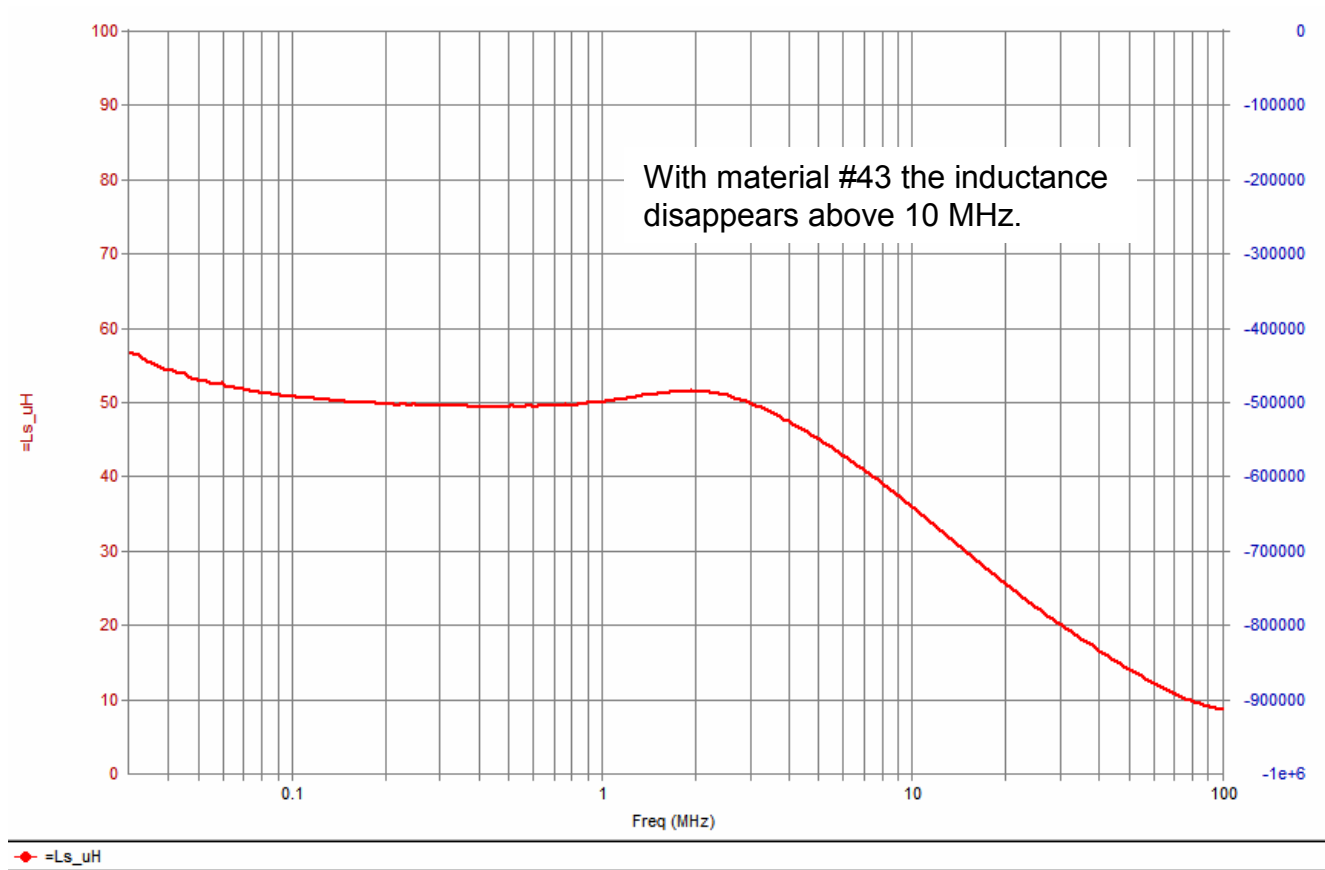
The Final Probe

Thanks Bert!

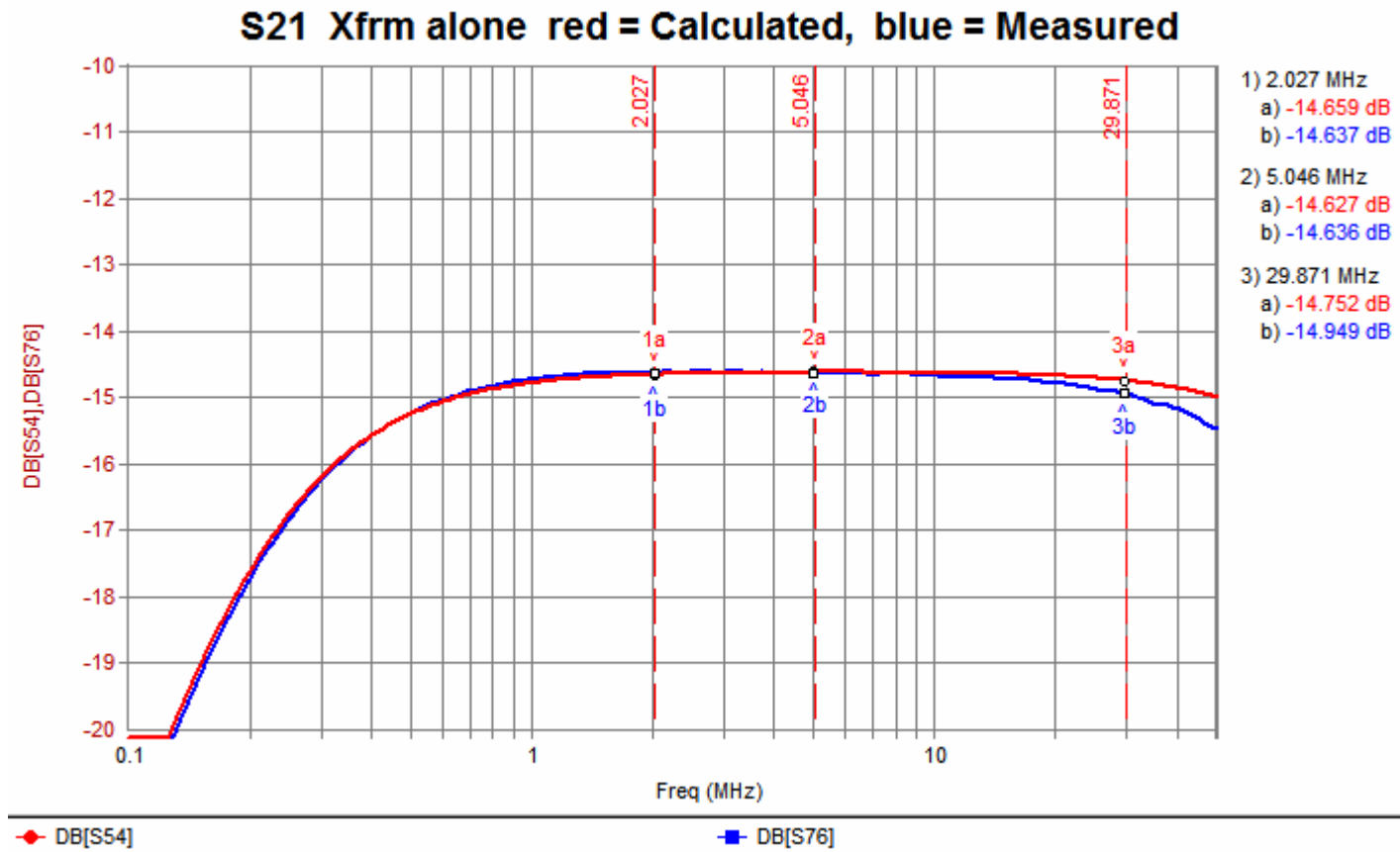


Measured inductance

11 turn coil on snap-on ferrite (10 turn was used to get $Z_{tr} > 5$)

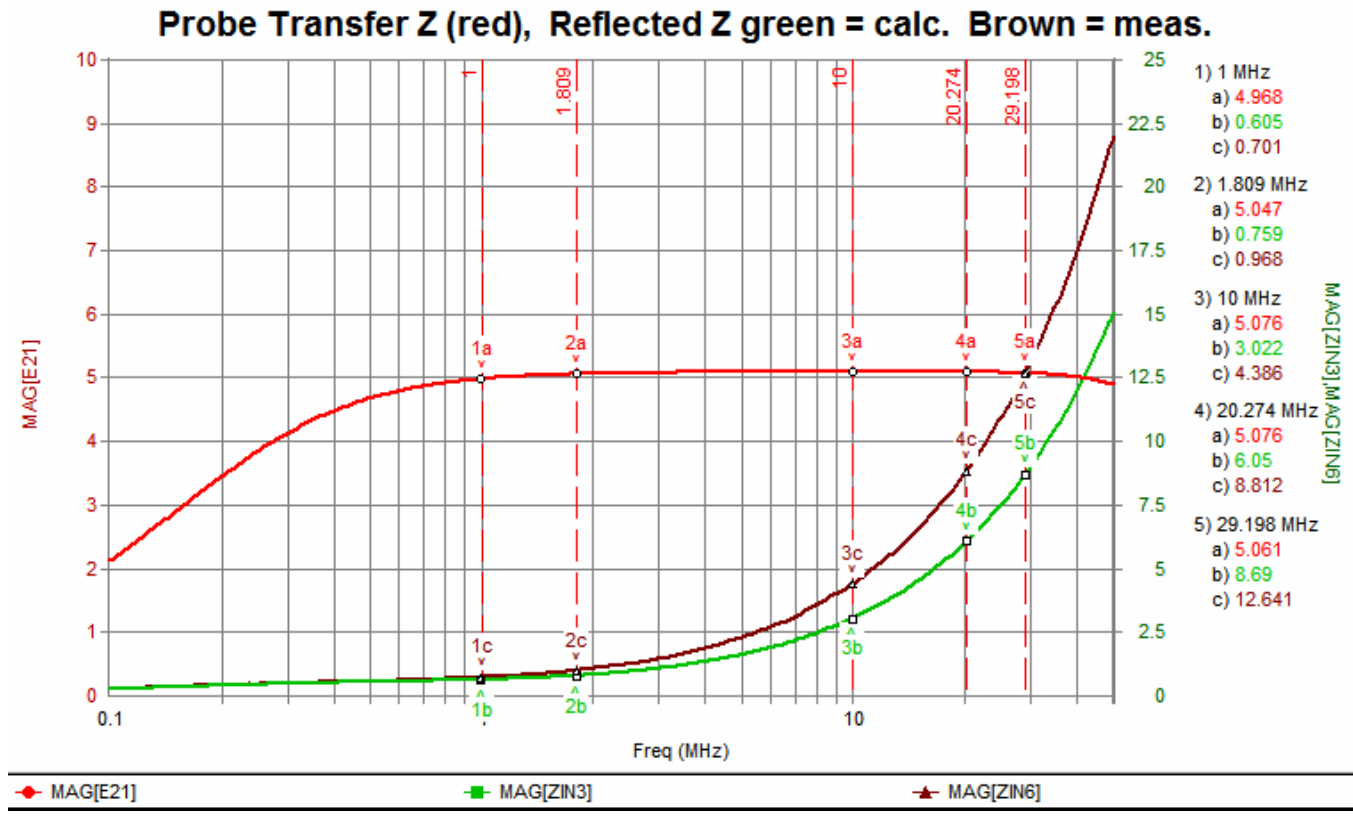


10 turn probe S21



During simulation: the coupling coefficient K is adjusted to get equal S21 at 5 MHz for both measured and calculated conditions. $K = 0.938$
The secondary inductance $L2$ was adjusted to obtain equal low freq. rolloff. Flow = 200 KHz
 $L2$ secondary = 39.5 μ H

10 turn probe S21



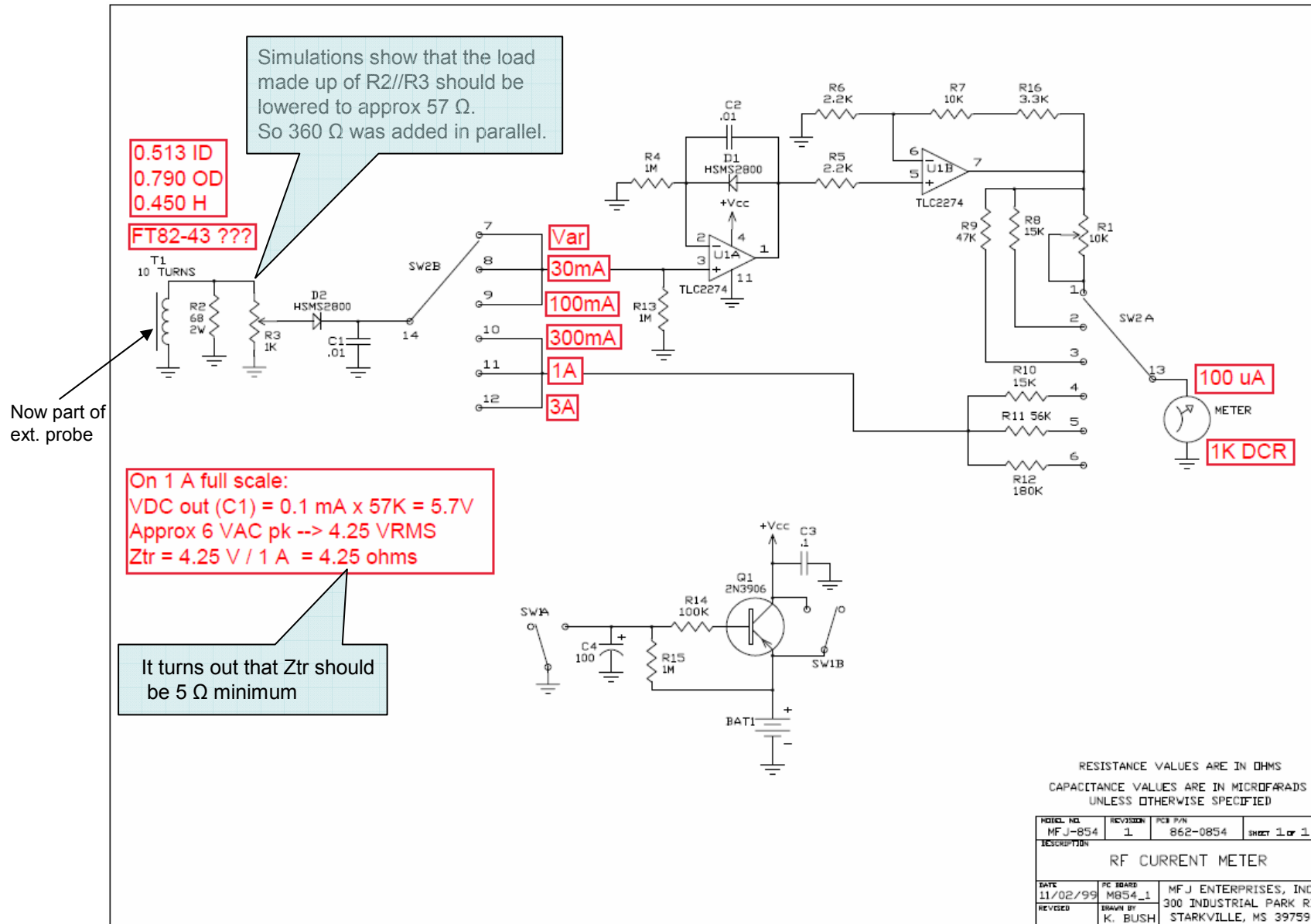
Probe transfer impedance (Ztr) is 5.07 Volts / Amps at 10 MHz (red curve) with the internal load ZL = 54.2 Ω, inside the MFJ-824.

Ztr at midband:
N = number of turns

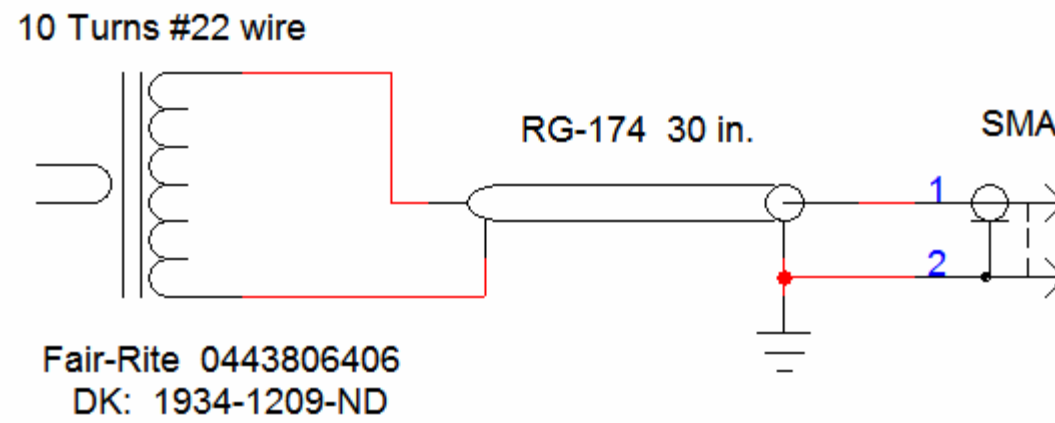
$$Z_{tr} = \frac{K * Z_L}{N} = 5.08$$

Probe sensitivity In
volts / amp for the MFJ-824
With ZL = 50 Ω Ztr = 4.7

Here K = 0.938
ZL = 54.2 Ω
N = 10



Schematic of external probe for MFJ-824



Calculations

$$F_c := \frac{Z_L}{2 \cdot \pi \cdot L_2} = 2.015 \times 10^5$$

Low Freq. Cutoff: F_c

$$R_r := \frac{K^2 \cdot L_1 \cdot Z_L}{L_2} = 0.44$$

Reflected Resistance above F_c

$$L_{eq} := L_1 \cdot (1 - K^2) = 4.746 \times 10^{-8}$$

Reflected inductance above F_c

$$Z_{tr}(K, N) := \frac{K \cdot Z_L}{N}$$

Z_{tr} is proportionnal to the K factor when $f \gg F_c$
(see below)

$$Z_{tr}(K, N) = 4.69$$

In Volts / Amp

$$M := K \cdot \sqrt{L_1 \cdot L_2}$$

$$M = 3.705 \times 10^{-6}$$

Mutual inductance

Power dissipated in the reflected resistance at 1 Amp input when $f \gg F_c$

$$P_{Rr} := \left(\frac{I_{in}}{N} \right)^2 \cdot \frac{K^2 \cdot L_1 \cdot Z_L}{L_2} = 4.399 \times 10^{-3}$$

Power dissipated in the Load resistance Z_L at 1 Amp input when $f \gg F_c$

$$P_{Rr} := \left(\frac{I_{in}}{N} \right)^2 \cdot Z_L = 0.5$$